# Technion—Israel Intitute of Technology The Erna and Andrew Viterbi Department of Electrical Engineering

Estimation and Identification in Dynamical Systems (048825) Lecture Notes: Prof. Nahum Shimkin, Spring 2016

## 5 The Continuous-Time Kalman Filter

The Model: Continuous-time linear system, with white noises state and measurement noises (not necessarily Gaussian).

**Goal:** Develop the continuous-time Kalman filter as the optimal *linear* estimator (L-MMSE) for this system.

One way to develop the continuous-time filter is as the limit (with  $\Delta T \to 0$ ) of the discrete time case. The derivation below follows a direct approach, based on the innovations process, introduced by Kailath, and will be somewhat informal.

A rigorous but very accessible treatment may be found in: M. Davis, *Linear Estimation and Stochastic Control*, 1977. However, this more advanced treatment is only essential in the nonlinear estimation problem.

#### 5.1 The continuous time model

We consider the state-space model

$$\frac{d}{dt}x_t = F_t x_t + w_t, \quad t \ge 0$$
$$z_t = H_t x_t + v_t$$

where:

•  $\{w_t\}$  and  $\{v_t\}$  are zero-mean white-noise processes, namely

$$E(w_t w_s^T) = Q_t \delta(t - s), \quad E(v_t v_s^T) = R_t \delta(t - s)$$
  
$$E(w_t v_s^T) = 0$$

- $\{w_t\}$ ,  $\{v_t\}$  and  $x_0$  are uncorrelated.
- $E(x_0) = \overline{x}_0$  and  $cov(x_0) = P_0$  are given.
- We shall assume that  $R_t$  is non-singular.

A simplifying assumption: We assume in the derivation below that  $\overline{x}_0 = 0$ , hence  $\overline{x}_t \doteq E(x_t) = 0$ . Otherwise,  $\overline{x}_t$  is given by  $\dot{\overline{x}}_t = F_t \overline{x}_t$  and needs to be added in some of the intermediate equations. The filter equations are the same.

Remark: The white processes above are not rigorously defined, due to the  $\delta$ -covariances, and indeed their sample-paths are quite "hectic". A rigorous definition of such processes (and the above model) is based on their integral — e.g., a Brownian motion in the Gaussian case.

State Covariance Propagation:  $\Pi_t := cov(x_t)$  satisfies

$$\frac{d}{dt}\Pi_t = F_t\Pi_t + \Pi_t F_t^T + Q_t \tag{1}$$

with  $\Pi_0 = \operatorname{cov}(x_0)$ .

*Proof:* A naive proof approach by differentiating  $E(x_t x_t^T)$  inside the expectation runs into trouble, because of the unusual properties of  $w_t$ . The following two options lead to the correct answer:

- 1. As a limit of the discrete-time case, with the approximation:  $x_{(k+1)\epsilon} = (I + \epsilon F_{k\epsilon})x_{k\epsilon} + w_{k\epsilon}$ , with  $E(w_{k\epsilon}w_{l\epsilon}) = \epsilon Q_{k\epsilon}\delta_{kl}$ , and  $\epsilon \to 0$ .
- 2. Explicitly solving for  $x_t$ : Let  $\Phi(t, s)$  denote the  $n \times n$  state transition matrix, namely the unique solution (for each s) of

$$\frac{d}{dt}\Phi(t,s) = F_t\Phi(t,s), \quad \Phi(s,s) = I \tag{2}$$

Then

$$x_t = \Phi(t,0)x_0 + \int_0^t \Phi(t,s)w_s ds$$
. (3)

This can be used to derive the covariance equation.

Example: Consider the stationary case – the system and covariance matrices are time independent.

Then the white noise processes have constant spectral densities:

$$S_w(\omega) = Q, \quad S_v(\omega) = R$$

and the noise-to state transfer function is

$$T(s) := T_{w \to x}(s) = (sI - F)^{-1}$$
.

When this system is stable, the state spectral density is given by  $S_x(\omega) = T(j\omega)QT^*(j\omega)$ , and the measurement spectral density is  $S_z(\omega) = HS_x(\omega)H^T + R$ .

## 5.2 Filter Derivation

Let  $Z_t = \{z_s, \ s < t\}$ . We need to calculate  $\hat{x}_t = E^L(x_t|Z_t)$ .

Define the innovations process:

$$\tilde{z}_t := z_t - E^L(z_t|Z_t) \tag{4}$$

Observe that

$$\tilde{z}_t = z_t - H_t E^L(x_t | Z_t) = H_t \tilde{x}_t + v_t \tag{5}$$

where  $\tilde{x}_t = x_t - \hat{x}_t$ .

## Properties of $\tilde{z}_t$ :

 $\tilde{z}_t$  is a zero-mean white noise process (exercise). Its covariance equals:

$$E(\tilde{z}_t \tilde{z}_s^T) = R_t \delta(t - s). \tag{6}$$

Note: same covariance as  $z_t$ !

It can also be shown that  $\tilde{Z}_t$  and  $Z_t$  are linearly equivalent, so that

$$E^L(\cdot|Z_t) = E^L(\cdot|\tilde{Z}_t)$$
.

It follows that  $\hat{x}_t$  can be expressed as a linear function of  $\tilde{Z}_t$ :

$$\hat{x}_t = \int_0^t g(t, s) \tilde{z}_s ds \tag{7}$$

The kernel g(t, s) is easily computable via the orthogonality principle. Since  $\tilde{x}_t := (x_t - \hat{x}_t) \perp \tilde{z}_s$  for s < t,

$$E(x_t \tilde{z}_s^T) = \int_0^t g(t, r) E(\tilde{z}_r \tilde{z}_s^T) dr$$
$$= \int_0^t g(t, r) \delta(s - r) R_s dr = g(t, s) R_s, \quad s < t.$$

Therefore,

$$\hat{x}_t = \int_0^t E(x_t \tilde{z}_s^T) R_s^{-1} \tilde{z}_s ds \tag{8}$$

Differentiate to obtain a differential equation for  $\hat{x}$ :

$$\frac{d}{dt}\hat{x}_t = \int_0^t E(\dot{x}_t \tilde{z}_s^T) R_s^{-1} \tilde{z}_s ds + E(x_t \tilde{z}_t^T) R_t^{-1} \tilde{z}_t \tag{9}$$

Now, define  $K_t = E(x_t \tilde{z}_t^T) R_t^{-1}$ , substitute  $\dot{x}_t$  from the state equation, note (8), and that  $E(w_t \tilde{z}_s^T) = 0$  for s < t. These give

$$\frac{d}{dt}\hat{x}_t = F_t\hat{x}_t + K_t\tilde{z}_t = F_t\hat{x}_t + K_t(z_t - H_t\hat{x}_t). \tag{10}$$

Using (5)

$$K_t := E(x_t \tilde{z}_t^T) R_t^{-1} = E(x_t \tilde{x}_t^T) H_t^T R_t^{-1} + 0 = P_t H_t^T R_t^{-1}$$
(11)

where  $P_t := E(\tilde{x}_t \tilde{x}_t^T)$ .

#### Calculating $P_t$ :

Recall that  $\tilde{x}_t = x_t - \hat{x}_t$ . Using (10),

$$\frac{d}{dt}\tilde{x}_t = (F_t - K_t H_t)\tilde{x}_t + w_t - K_t v_t. \tag{12}$$

As in (1), this implies

$$\frac{d}{dt}P_t = (F_t - K_t H_t)P_t + P_t (F_t - K_t H_t)^T + Q_t + K_t R_t K_t^T$$
(13)

with  $P_0$  given.

An alternative expression: by substituting  $K_t$  from (11) and rearranging,

$$\frac{d}{dt}P_{t} = F_{t}P_{t} + P_{t}F_{t}^{T} + Q_{t} - K_{t}R_{t}K_{t}^{T}.$$
(14)

This is the (differential) Riccati Equation - a quadratic matrix differential equation.

To summarize: The filter equation is given by (10)

$$\frac{d}{dt}\hat{x}_t = F_t\hat{x}_t + K_t\tilde{z}_t = F_t\hat{x}_t + K_t(z_t - H_t\hat{x}_t)$$

with the gain (11)

$$K_t := E(x_t \tilde{z}_t^T) R_t^{-1} = P_t H_t^T R_t^{-1}.$$

The covariance may be computed by (the Joseph form)

$$\frac{d}{dt}P_{t} = (F_{t} - K_{t}H_{t})P_{t} + P_{t}(F_{t} - K_{t}H_{t})^{T} + Q_{t} + K_{t}R_{t}K_{t}^{T}$$

or (14)

$$\frac{d}{dt}P_t = F_t P_t + P_t F_t^T + Q_t - K_t R_t K_t^T.$$

Note: if the state and measurement noises are correlated, namely

$$E(w_t v_s^T) = S_t \delta(t - s) ,$$

then the gain in (11) becomes  $K_t = (P_t H_t^T + S_t) R_t^{-1}$ , and the covariance update (13) should be modified by adding  $-(S_t K_t^T + K_t S_t^T)$  on the right.