Patch-Ordering-Based Wavelet Frame and Its Use in Inverse Problems

Idan Ram, Israel Cohen, Senior Member, IEEE, and Michael Elad, Fellow, IEEE

Abstract-In our previous work [1] we have introduced a redundant tree-based wavelet transform (RTBWT), originally designed to represent functions defined on high dimensional data clouds and graphs. We have further shown that RTBWT can be used as a highly effective image-adaptive redundant transform that operates on an image using orderings of its overlapped patches. The resulting transform is robust to corruptions in the image, and thus able to efficiently represent the unknown target image even when it is calculated from its corrupted version. In this paper, we utilize this redundant transform as a powerful sparsity-promoting regularizer in inverse problems in image processing. We show that the image representation obtained with this transform is a frame expansion, and derive the analysis and synthesis operators associated with it. We explore the use of this frame operators to image denoising and deblurring, and demonstrate in both these cases state-of-the-art results.

Index Terms—Patch-based processing, redundant wavelet, frames, denoising, deblurring, ordering, regularization.

I. INTRODUCTION

S PARSE and redundant representations and the processing of local patches have become two of the most popular approaches in image processing in recent years. While image processing algorithms may be based only on patch processing [2], [3] or sparse representations [4], [5], many current stateof-the-art algorithms make use of both concepts, usually by processing the image patches using sparsifying transforms or learned dictionaries [6]–[11].

In our previous work [1] and [12] we have combined the two aforementioned approaches in a different manner, and used image patches to construct both an orthogonal and a redundant wavelet transforms, which efficiently (sparsely) represent entire images. These two wavelet transforms were originally designed to represent scalar functions defined on high-dimensional data clouds and graphs. However, we have also shown in [1] and [12] that the very same construction can be used as an image-adaptive transform that is highly

Manuscript received November 22, 2013; revised April 13, 2014; accepted April 17, 2014. Date of publication April 23, 2014; date of current version May 20, 2014. This work was supported in part by the European Research Council through the European Union's Seventh Framework Program under Grant 320649, in part by Robert H. Hillman Foundation for Global Security through the Collaboration Technion and University Northeastern, in part by the Israel Science Foundation under Grant 1130/11, and in part by the Japan Technion Society Research Fund. The associate editor coordinating the review of this manuscript and approving it for publication was Prof. Chang-Su Kim.

I. Ram and I. Cohen are with the Department of Electrical Engineering, Technion–Israel Institute of Technology, Haifa 32000, Israel (e-mail: idanram@tx.technion.ac.il; icohen@ee.technion.ac.il).

M. Elad is with the Department of Computer Science, Technion–Israel Institute of Technology, Haifa 32000, Israel (e-mail: elad@cs.technion.ac.il). Color versions of one or more of the figures in this paper are available online at http://ieeexplore.ieee.org.

Digital Object Identifier 10.1109/TIP.2014.2319739

effective for sparsifying image content. This is obtained by converting the given image into a graph by considering all its overlapped patches as coordinates in high-dimensional space, and referring to them as features of the graph vertices. These are accompanied by their mutual Euclidean distance to define the graph edges, this way tying the vertices to each other. Due to the reliance on patches, both transforms are robust to corruptions in the image, such as additive noise, blur, or missing values, and are able to efficiently represent the unknown target image even when they are calculated from its corrupted version. As we shall see hereafter, this work will utilize this last property, and demonstrate the use of our redundant transform proposed in [1], (termed redundant treebased wavelet transform – RTBWT), as a powerful sparsitypromoting regularizer in inverse problems.

More specifically, the RTBWT is calculated for an image by adding data-dependent operators, merged into the classical redundant wavelet filter-bank implementation [13], [14]. In each decomposition level several operators are used to reorder the approximation coefficients before the scaling and wavelet filters are applied to them. The reordering operators are obtained by organizing feature vectors constructed from the image patches, such that they are chained in the shortest possible path, essentially obtaining an approximation to the solution of the traveling salesman problem (TSP) [15]. These permutation operators increase the regularity of the approximation coefficient signals, thus causing their representation with the RTBWT to be more efficient (sparse).

As said above, in this work we utilize the RTBWT as a powerful sparsity-promoting regularizer in inverse problems in image processing. We start by introducing a simpler implementation to the RTBWT, which is based on the widely known "à trous" algorithm [16], [17]. This algorithm is a different implementation of the redundant wavelet transform that applies in each decomposition level upsampled versions of the scaling and wavelet filters to the whole approximation coefficient vectors. Thus, our scheme essentially adds to each decomposition level of this transform a single permutation operator that reorders the approximation coefficients before they are filtered, leading to the "à trous" implementation of our transform. We use this scheme and ideas from [18] to show that the RTBWT is a valid frame expansion, with the same bounds as the common redundant wavelet transform.

In our previous works [1], [3], and [12] we observed that the performance of patch-ordering-based algorithms improve when a subimage averaging scheme is used. For algorithms using the RTBWT, this consists of applying the transform to different subimages of the treated image and then jointly

1057-7149 © 2014 IEEE. Personal use is permitted, but republication/redistribution requires IEEE permission. See http://www.ieee.org/publications_standards/publications/rights/index.html for more information.



Fig. 1. RTBWT decomposition scheme.

operating on the transform coefficients of all the subimages. We refer to all the transform coefficients of the different subimages as a single extended representation of the treated image, and next take a path similar to the one described in [11] for the BM3D algorithm. We construct matrices which act as analysis and synthesis operators [19], and are used to obtain this extended representation and reconstruct the image from it. We then show that these matrices constitute a frame and its dual. We explore the use of these operators in image denoising and deblurring algorithms. Despite the fact that the resulting transform is image dependent, we treat it as a fixed linear operator and use it within a sparsity-promoting regularizer, when handling image-processing tasks posed as inverse problems. We demonstrate state-of-the-art results in denoising and deblurring using this approach.

The paper is organized as follows: In Section II, we describe the image-derived redundant tree-based wavelet transform. We also introduce the à trous based implementation of the transform, and use it to analyze its frame properties. Section III introduces the RTBWT subimage averaging frame, and analyzes its properties. Sections II and III can be skipped by readers interested in the image processing applications side of this work. In Section IV, we explore the use of this frame to image denoising and deblurring, and present experimental results that demonstrate their advantages. We summarize the paper in Section V.

II. REDUNDANT TREE-BASED WAVELET TRANSFORM

This and the next sections are dedicated to a careful construction and study of the frames that will be later used for regularizing inverse problems.

A. Decomposition and Reconstruction Schemes

Let **Y** be an image of size $N_1 \times N_2$ where $N_1N_2 = N$, and let **y** be the column stacked version of **Y**. The redundant tree-based wavelet transform (RTBWT), introduced in [1], is designed to efficiently (sparsely) represent its input vector, which in our case is **y**. The transform is constructed by modifying an implementation of the redundant wavelet transform proposed by Shensa [13] and Beylkin [14]. Figure 1 describes the decomposition scheme of the RTBWT. The filters \mathbf{h} and \mathbf{g} are the scaling and wavelet decomposition filters of an orthonormal discrete wavelet transform, and they are applied using cyclic convolution. The 2 : 1 decimators denoted by

| Algorithm 1 RTBWT L-Level Decomposition Scheme |
|---|
| Task: Apply on y an L-level RTBWT decomposition. |
| Parameters: The image y, the operators $\{\mathbf{P}_{\ell}^{s}\}_{s=1}^{2^{\ell}}$, |
| $\ell = 0, \dots, L-1$, and the filters h and g . |
| Initialization: Set $\mathbf{a}_0^1 = \mathbf{a}_0 = \mathbf{y}$. |
| Main Iteration: Perform the following steps for |
| $\ell = 0, \dots, L-1$: |
| • Perform the following steps for $s = 1,, 2^{\ell}$: |
| - Apply \mathbf{P}_{ℓ}^{s} on \mathbf{a}_{ℓ}^{s} and obtain $\mathbf{a}_{\ell}^{s,p}$. |
| - Filter $\mathbf{a}_{\ell}^{s,p}$ with $\bar{\mathbf{g}}$ and obtain $\mathbf{d}_{\ell+1}^{s}$. |
| – Filter $\mathbf{a}_{\ell}^{s,p}$ with $ar{\mathbf{h}}$ and decimate the result with |
| * $\downarrow 2^o$ and obtain $\mathbf{a}_{\ell+1}^s$. |
| $* \hspace{0.1 in} \downarrow 2^{e}$ and obtain $\mathbf{a}_{\ell+1}^{s+2^{\ell}}$ |
| Output: The approximation coefficient vectors \mathbf{a}_L and |
| detail coefficient vectors $\{\mathbf{d}_{\ell}\}_{\ell=1}^{L}$ |

 $\downarrow 2^o$ and $\downarrow 2^e$ keep the odd and even samples of their input, respectively. The signals \mathbf{a}_{ℓ} and \mathbf{d}_{ℓ} are the approximation and detail coefficient vectors in the ℓ th scale, respectively, where $\mathbf{a}_0 = \mathbf{y}$. We note that for $\ell > 1$ these signals do not appear explicitly in the decomposition scheme, and instead it employs the signals denoted by \mathbf{a}_{ℓ}^s and \mathbf{d}_{ℓ}^s , which contain subsets of the samples in \mathbf{a}_{ℓ} and \mathbf{d}_{ℓ} , respectively. The signal \mathbf{a}_{ℓ}^s is obtained from \mathbf{a}_{ℓ} by starting from the *s*th sample, and keeping every 2^{ℓ} th sample. The signal \mathbf{d}_{ℓ}^s is obtained from \mathbf{d}_{ℓ} by starting from the *s*th sample, and keeping every $2^{\ell-1}$ th sample. For example, for signals \mathbf{a}_2 and \mathbf{d}_2 of length 16 we get that

$$\mathbf{a}_{2}^{1} = [a_{2}[1], a_{2}[5], a_{2}[9], a_{2}[13]]^{T}$$

$$\mathbf{a}_{2}^{2} = [a_{2}[2], a_{2}[6], a_{2}[10], a_{2}[14]]^{T}$$

$$\mathbf{a}_{2}^{3} = [a_{2}[3], a_{2}[7], a_{2}[11], a_{2}[15]]^{T}$$

$$\mathbf{a}_{2}^{4} = [a_{2}[4], a_{2}[8], a_{2}[12], a_{2}[16]]^{T}$$

$$\mathbf{d}_{2}^{1} = [d_{2}[1], d_{2}[3], \dots, d_{2}[15]]^{T}$$

$$\mathbf{d}_{2}^{2} = [d_{2}[2], d_{2}[4], \dots, d_{2}[16]]^{T}.$$
(1)

We notice that $\mathbf{a}_0^1 = \mathbf{a}_0$ and $\mathbf{d}_1^1 = \mathbf{d}_1$.

The operators \mathbf{P}_{ℓ}^{s} make the difference between our proposed wavelet decomposition scheme and the common redundant wavelet transform [13], [14]. Each such operator produces a permuted version $\mathbf{a}_{\ell}^{s,p}$ of its input vector \mathbf{a}_{ℓ}^{s} , and it may be interpreted as a linear and unitary operator given that vector. These operators "smooth" the approximation coefficient signals in the different levels of the decomposition scheme. In Section II-B we explain how to obtain these operators from the image patches. Assuming that the operators \mathbf{P}_{ℓ}^{s} are known, Algorithm 1 is used to apply an *L*-level RTBWT decomposition.

In a similar manner, Figure 2 describes the reconstruction scheme of the redundant tree-based wavelet transform. If **h** and **g** denote the scaling and wavelet reconstruction filters, then $\tilde{\mathbf{h}} = \frac{1}{2}\mathbf{h}$ and $\tilde{\mathbf{g}} = \frac{1}{2}\mathbf{g}$ are applied using cyclic convolution. The upsamplers denoted by $\uparrow 2^o$ and $\uparrow 2^e$ place the samples of their input vector in the odd and even locations of their output vector, respectively. The operator $\tilde{\mathbf{P}}_{\ell}^s$ reorders a vector so as to cancel the ordering done by \mathbf{P}_{ℓ}^s , i.e. $\tilde{\mathbf{P}}_{\ell}^s = (\mathbf{P}_{\ell}^s)^{-1} = (\mathbf{P}_{\ell}^s)^T$. Assuming that the operators $\tilde{\mathbf{P}}_{\ell}^s$ are known, Algorithm 2



Fig. 2. RTBWT reconstruction scheme.

reconstructs **y** from an *L*-level RTBWT decomposition. We next explain how the operators \mathbf{P}_{ℓ}^{s} are determined in each level of the RTBWT.

B. Building the Operators \mathbf{P}^{s}_{ℓ}

We wish to design the operators \mathbf{P}_{ℓ}^{s} in a manner which results in an efficient (sparse) representation of the input image by the proposed transform. The wavelet transform is known to produce a small number of large coefficients when it is applied to piecewise regular signals [16]. Thus, we would like the operator \mathbf{P}_{ℓ}^{s} , applied to \mathbf{a}_{ℓ}^{s} , to produce a signal $\mathbf{a}_{\ell}^{s,p}$ which is as regular as possible. We start with the finest level, and try to find the permutation that the operator \mathbf{P}_{0}^{1} applies to $\mathbf{a}_{0} = \mathbf{y}$. When the image \mathbf{Y} is known, the optimal solution would be to apply a simple sort operation on \mathbf{y} . However, since we are interested in the case where \mathbf{y} may be corrupted (noisy, blurred, contain missing pixels, etc.), we would try to find a near-optimal ordering operation using the image patches.

Let y_i denote the *i*th sample in the vector \mathbf{y} , and let \mathbf{x}_i denote the column stacked version of the $\sqrt{n} \times \sqrt{n}$ patch around the location of y_i in \mathbf{Y} . A key assumption in our work is that under some distance measure $w(\mathbf{x}_i, \mathbf{x}_j)$, proximity between the two patches \mathbf{x}_i and \mathbf{x}_j suggests proximity between their center pixels y_i and y_j . Thus, we shall try to reorder the patches \mathbf{x}_i so that they form a smooth path, hoping that the corresponding reordered 1D signal $\mathbf{a}_0^{1,p} = \mathbf{y}^p$ will also be smooth. The "smoothness" of the reordered signal $\mathbf{a}_0^{1,p}$ can be measured using its total variation measure

$$\|\mathbf{a}_{0}^{1,p}\|_{TV} = \sum_{j=2}^{N} |a_{0}^{1,p}[j] - a_{0}^{1,p}[j-1]|.$$
(2)

Let $\{\mathbf{x}_{j}^{p}\}_{j=1}^{N}$ denote the patches $\{\mathbf{x}_{i}\}_{i=1}^{N}$ in their new order. Then by analogy, we evaluate the "smoothness" of the path through the patches \mathbf{x}_{i}^{p} by the measure

$$TV(\mathbf{x}_{j}^{p}) = \sum_{j=2}^{N} w(\mathbf{x}_{j}^{p}, \mathbf{x}_{j-1}^{p}).$$
(3)

Minimizing $TV(\mathbf{x}_j^p)$ comes down to finding the shortest path that passes through the set of points \mathbf{x}_i , visiting each point only once. This can be regarded as an instance of the traveling salesman problem (TSP) [15], which can become



– Apply
$$\tilde{\mathbf{P}}_{\ell}^{s}$$
 on $\mathbf{a}_{\ell}^{s,p}$ and obtain \mathbf{a}_{ℓ}^{s} .

Output: The reconstructed signal $\mathbf{a}_0 = \mathbf{y}$.

very computationally exhaustive for large sets of points. We choose a simple and crude approximation to the solution, which is to start from an arbitrary point (random or predefined), and continue from each point to its nearest neighbor, not visiting any point twice. Let \mathbf{q}_i denote a vector containing the 2D spatial coordinate of the patch \mathbf{x}_i in the image Y. We restrict the nearest neighbor search performed for each patch \mathbf{x}_i to a square neighborhood of size $B \times B$ around \mathbf{q}_i . When no unvisited patch remains in that neighborhood, we search for nearest neighbors among all the unvisited patches in the whole image. This restriction decreases the overall computational complexity, and our experiments show that with a proper choice of B (found empirically so as to optimize the results) it also leads to improved results, as it forces more relevant neighbors in the ordering. Further work is needed to set this parameter automatically and perhaps even consider a dynamic search area based on the patch-content. The permutation applied by the operator \mathbf{P}_0^1 is defined as the order of the found path.

In order to further increase the sparseness of the image representation with RTBWT, we use the aforementioned method to find operators $\{\mathbf{P}_{\ell}^{s}\}_{s=1}^{2^{\ell}}$ which are applied to the signals $\{\mathbf{a}_{\ell}^{s}\}_{s=1}^{2^{\ell}}$ in scales $\ell > 0$. To this end, we are required to associate a set of feature points with these signals. More specifically, we predict the proximity between the samples of the signal \mathbf{a}_{ℓ}^{s} , by associating a feature point $\mathbf{x}_{\ell,i}^{s}$ with each sample $a_{\ell}^{s}(j)$. Also, in order to measure the spatial proximity between the feature points, we associate a 2D spatial coordinate $\mathbf{q}_{\ell i}^{s}$ to each feature point $\mathbf{x}_{\ell i}^{s}$. The calculation of the feature points and their coordinates is carried out in a recursive manner. We use the set of ℓ -th scale feature points $\{\mathbf{x}_{\ell,j}^s\}_{j=1}^{2^{-\ell}N}$ to calculate the two sets of feature points $\{\mathbf{x}_{\ell+1,j}^s\}_{j=1}^{2^{-\ell-1}N}$ and $\{\mathbf{x}_{\ell+1,j}^{s+2^{\ell}}\}_{j=1}^{2^{-\ell-1}N}$, used in the $\ell + 1$ th scale. This is done in analogy to the way we obtain the signals $\mathbf{a}_{\ell+1}^{s}$ and $\mathbf{a}_{\ell+1}^{s+2^{\ell}}$ from the signal \mathbf{a}_{ℓ}^{s} . We first order the feature points $\{\mathbf{x}_{\ell,j}^{s}\}_{j=1}^{2^{-\ell}N}$ according to the permutation defined by \mathbf{P}_{ℓ}^{s} , and place them in the columns of an $n \times 2^{-\ell} N$ matrix $\mathbf{X}_{\ell}^{s,p}$. We then filter the rows of the result with $\bar{\mathbf{h}}^T$, and obtain the

sets of feature points $\{\mathbf{x}_{\ell+1,j}^s\}_{j=1}^{2^{-\ell-1}N}$ and $\{\mathbf{x}_{\ell+1,j}^{s+2^\ell}\}_{j=1}^{2^{-\ell-1}N}$ by keeping only the odd and even columns of the resulting matrix, respectively.

A similar process is used to calculate the two sets of spatial coordinates $\{\mathbf{q}_{\ell+1,j}^s\}_{j=1}^{2^{-\ell-1}N}$ and $\{\mathbf{q}_{\ell+1,j}^{s+2^{\ell}}\}_{j=1}^{2^{-\ell-1}N}$ corresponding to $\{\mathbf{x}_{\ell+1,j}^s\}_{j=1}^{2^{-\ell-1}N}$ and $\{\mathbf{x}_{\ell+1,j}^{s+2^{\ell}}\}_{j=1}^{2^{-\ell-1}N}$, respectively. However, here we filter the rows of the result with a normalized filter \mathbf{h}_a^T , which satisfies

$$\bar{h}_a[n] = |\bar{h}[n]| / \sum_k |\bar{h}[k]|,$$
(4)

so as to guarantee no drift in the coordinates.

We note that every time we advance from one scale to the next, the number of features points used to calculate a single reordering operator decreases by a factor of two. Therefore, when we move to the coarser level we increase both dimensions of the search area by a factor of $\sqrt{2}$, so as to keep the number of candidate neighbors to consider. Thus, the size of the search area used in the ℓ th scale is $2^{\frac{\ell}{2}}B \times 2^{\frac{\ell}{2}}B$.

Having calculated the feature points $\mathbf{x}_{\ell+1,j}^s$ and their coordinates $\mathbf{q}_{\ell+1,i}^{s}$, they can be fed to the approximate shortest path search method described above to obtain each of the operators $\mathbf{P}_{\ell+1}^{s}$ in the ℓ + 1th scale. The calculation scheme of all the ordering operators \mathbf{P}_{ℓ}^{s} that are used in an L-level RTBWT decomposition is summarized in Algorithm 3. Figure 3 shows a scheme describing an example for the such calculation for a 2-level decomposition of an image with 8 pixels. We note that Similarly to the decomposition scheme of the generalized treebased wavelet transform (GTBWT) described in [12], the relation between the feature points in a full decomposition can be described using tree-like structures. Each such "generalized" tree contains all the feature points which have participated in the calculation of a single feature point $\mathbf{x}_{L,i}^{s}$ from the coarsest scale. Figure 4 shows an example of such a "generalized" tree, corresponding to length N = 8, using a filter **h** of length 4 and disregarding boundary issues in the different levels. We next describe an alternative implementation of the RTBWT, and use it to show the frame properties of the transform.

C. À Trous Implementation and Frame Properties

The common and widely known "à trous" algorithm [16], [17] is an alternative implementation for the redundant wavelet transform. Instead of explicitly applying the filters $\bar{\mathbf{h}}$ and $\bar{\mathbf{g}}$ to subsets of the signals \mathbf{a}_{ℓ} , the à trous algorithm applies upsampled versions of these filters directly to \mathbf{a}_{ℓ} . The upsampled filters are obtained by inserting "holes" (trous in French) between nonzero filter taps. We next show that the RTBWT can also be applied using a modification of the à trous algorithm. We then use this scheme to show that the RTBWT is a frame expansion, with the same bounds as the common redundant wavelet transform. We note that the à trous implementation and the frame properties of the RTBWT are valid in general, and are not restricted to the image-derived RTBWT.

We first notice that in the aforementioned implementation of the RTBWT decomposition scheme, the operators \mathbf{P}_{ℓ}^{s} operate in the ℓ th scale on disjoint subvectors \mathbf{a}_{ℓ}^{s} of \mathbf{a}_{ℓ} . Therefore, these operators can be replaced by a single operator Algorithm 3 Calculation of All the Reordering Operators Used in an *L*-Level RTBWT Decomposition

Task: Calculate all the operators $\{\mathbf{P}_{\ell}^{s}\}_{s=1}^{2^{\ell}}, \ell = 0 \dots L - L$ 1, used in an *L*-level RTBWT decomposition. **Parameters:** The image patches $\{\mathbf{x}_j\}_{j=1}^N$, their 2D coordinates $\{\mathbf{q}_j\}_{j=1}^N$, the distance function w, the filters $\bar{\mathbf{h}}$ and $\bar{\mathbf{h}}_a$, and the search area size *B*. Initialization: Set $\mathbf{x}_{0,j}^1 = \mathbf{x}_j$ and $\mathbf{q}_{0,j}^1 = \mathbf{q}_j$. Main Iteration: Perform the following steps for $\ell = 0, \dots, L - 1$: • Set $\tilde{B} = 2^{\frac{\ell}{2}} B$ and $\tilde{N} = 2^{-\ell} N$. • Perform the following steps for $s = 1, \ldots, 2^{\ell}$: Calculate \mathbf{P}_{ℓ}^{s} : - Choose a random index $1 \leq j \leq \tilde{N}$ and set $\Omega(1) = \{j\}.$ Perform the following steps for k = 1, ..., N - 1: Set A_k to be the set of indices of the coordinates $\mathbf{q}_{\ell,i}^s$ which reside inside a $B \times B$ neighborhood around $\mathbf{q}_{\ell,\Omega(k)}^s$. If $|A_k \setminus \Omega| > 0$, find $\mathbf{x}_{\ell,\Omega_k}^s$ — the minimizer of $w(\mathbf{x}_{\ell,j}^s, \mathbf{x}_{\ell,\Omega(k)}^s)$ such that $j_1 \in A_k$ and $j_1 \notin \Omega$. Else, find \mathbf{x}_{ℓ,j_1}^s — the minimizer of $w(\mathbf{x}_{\ell,j_1}^s, \mathbf{x}_{\ell,\Omega(k)}^s)$ such that $j_1 \notin \Omega$.

Set
$$\Omega(k+1)$$
 to be j_1 .

- The set Ω holds the ordering applied by \mathbf{P}_{ℓ}^{s} .

Calculate $\ell + 1$ th scale feature points:

- Order the feature points $\{\mathbf{x}_{\ell,j}^s\}_{j=1}^{\tilde{N}}$ according to the permutation defined by \mathbf{P}_ℓ^s
- Place the reordered points in the columns of a matrix X^{s,p}_ℓ.
- Apply the filter $\bar{\mathbf{h}}^T$ to the rows of $\mathbf{X}_{\ell}^{s,p}$ and
 - * keep the odd columns of the result so as to obtain the points $\{\mathbf{x}_{i}^{s}, ..., \}^{\frac{\tilde{N}}{2}}$
 - obtain the points $\{\mathbf{x}_{\ell+1,j}^s\}_{j=1}^{\frac{N}{2}}$. * keep the even columns of the result so as to obtain the points $\{\mathbf{x}_{\ell+1,j}^{s+2^\ell}\}_{j=1}^{\frac{N}{2}}$.

Calculate $\ell + 1$ th scale coordinates:

- Order the coordinates $\{\mathbf{q}_{\ell,j}^s\}_{j=1}^{\tilde{N}}$ according to the permutation defined by \mathbf{P}_{ℓ}^s .
- Place the reordered points in the columns of a matrix Q^{s,p}_ℓ.
- Apply the filter $\bar{\mathbf{h}}_a^T$ to the rows of $\mathbf{Q}_\ell^{s,p}$ and
 - * keep the odd columns of the result so as to obtain the points $\{\alpha_{\ell+1,i}^s\}_{i=1}^{\frac{N}{2}}$.
 - obtain the points $\{\mathbf{q}_{\ell+1,j}^s\}_{j=1}^{\frac{N}{2}}$. * keep the even columns of the result so as to obtain the points $\{\mathbf{q}_{\ell+1,j}^{s+2^{\ell}}\}_{j=1}^{\frac{N}{2}}$.

Output: The operators $\{\mathbf{P}_{\ell}^s\}_{s=1}^{2^{\ell}}, \ell = 0, \dots, L-1.$

 \mathbf{P}_{ℓ} which operates on \mathbf{a}_{ℓ} , and produces the reordered signal \mathbf{a}_{ℓ}^{p} . Let \mathbf{R}_{ℓ}^{s} be a matrix which extracts the signal \mathbf{a}_{ℓ}^{s} from the signal \mathbf{a}_{ℓ} . Then we construct the permuted signal \mathbf{a}_{ℓ}^{p} by extracting all the signals $\{\mathbf{a}_{\ell}^{s}\}_{s=1}^{2\ell}$ from \mathbf{a}_{ℓ} , applying to each signal \mathbf{a}_{ℓ}^{s} the corresponding operator \mathbf{P}_{ℓ}^{s} , and returning all the permuted signals back to their original location. More specifically,

$$\mathbf{a}_{\ell}^{p} = \sum_{s=1}^{2^{\ell}} (\mathbf{R}_{\ell}^{s})^{T} \mathbf{P}_{\ell}^{s} \mathbf{R}_{\ell}^{s} \mathbf{a}_{\ell} = \mathbf{P}_{\ell} \mathbf{a}_{\ell}$$
(5)

where the matrices $(\mathbf{R}_{\ell}^{s})^{T}$ return the permuted signals back



Fig. 3. Calculation of the reordering operators for a 2-level decomposition of an image with 8 pixels.



Fig. 4. An illustration of a "generalized" tree.

to their locations. This way we have defined the permutation matrix

$$\mathbf{P}_{\ell} = \sum_{s=1}^{2^{\circ}} (\mathbf{R}_{\ell}^{s})^{T} \mathbf{P}_{\ell}^{s} \mathbf{R}_{\ell}^{s}.$$
 (6)

We can use the operators \mathbf{P}_{ℓ} to modify the à trous algorithm, and obtain the à trous implementation of the RTBWT decomposition scheme. Let $\bar{\mathbf{h}}_{\ell}$ be a filter obtained from $\bar{\mathbf{h}}$ by inserting $2^{\ell} - 1$ zeros between each sample, where $\bar{\mathbf{h}}_0 = \bar{\mathbf{h}}$. Also, let $\bar{\mathbf{g}}_{\ell}$ be a filter obtained from $\bar{\mathbf{g}}$ in a similar manner. Both filters are applied using cyclic convolution. Then assuming that the operators \mathbf{P}_{ℓ} are known, we can apply the RTBWT by repeating for $\ell = 0, \ldots, L - 1$ the following filter bank operations, described in Figure 5,

$$\mathbf{a}_{\ell}^{p} = \mathbf{P}_{\ell} \mathbf{a}_{\ell} \tag{7}$$



Fig. 5. RTBWT à trous based decomposition scheme.

$$\mathbf{a}_{2} \rightarrow \mathbf{b}_{1} \rightarrow \mathbf{v}_{1/2} \rightarrow \mathbf{\tilde{P}}_{1} \rightarrow \mathbf{a}_{1} \rightarrow \mathbf{b} \rightarrow \mathbf{v}_{1/2} \rightarrow \mathbf{\tilde{P}}_{0} \rightarrow \mathbf{a}_{0}$$
$$\mathbf{d}_{2} \rightarrow \mathbf{g}_{1} \qquad \mathbf{d}_{1} \rightarrow \mathbf{g} \rightarrow \mathbf{g}_{1} \rightarrow \mathbf{g}$$

Fig. 6. RTBWT à trous based reconstruction scheme.

$$a_{\ell+1}[n] = a_{\ell}^p * \bar{h}_{\ell}[n] \tag{8}$$

$$d_{\ell+1}[n] = a_{\ell}^{p} * \bar{g}_{\ell}[n].$$
(9)

We note that when the operators \mathbf{P}_{ℓ} are removed, our scheme coincides with the common à trous algorithm.

In a similar manner, the operators $\tilde{\mathbf{P}}_{\ell}^{s} = (\mathbf{P}_{\ell}^{s})^{-1}$, which operate on the signals $\mathbf{a}_{\ell}^{s,p}$ can also be replaced by a single operator $\tilde{\mathbf{P}}_{\ell} = (\mathbf{P}_{\ell})^{-1}$. We use the operators $\tilde{\mathbf{P}}_{\ell}$ to construct the à trous implementation of the RTBWT reconstruction scheme. Let \mathbf{h}_{ℓ} be a filter obtained from \mathbf{h} by inserting $2^{\ell} - 1$ zeros between each sample, where $\mathbf{h}_0 = \mathbf{h}$. Also, let \mathbf{g}_{ℓ} be a filter obtained from \mathbf{g} in a similar manner. Both filters are applied using cyclic convolution. Then assuming that the operators $\tilde{\mathbf{P}}_{\ell}$ are known, we can apply the RTBWT reconstruction scheme by repeating for $\ell = L - 1, \ldots, 0$ the following filter bank operations, described in Figure 6,

$$a_{\ell}^{p}[n] = \frac{1}{2} \left(a_{\ell+1} * h_{\ell}[n] + d_{\ell+1} * g_{\ell}[n] \right)$$
(10)

$$\mathbf{a}_{\ell} = \tilde{\mathbf{P}}_{\ell} \mathbf{a}_{\ell}^{p}. \tag{11}$$

D. Frame Properties

Let Φ denote an $(L + 1)N \times N$ transform matrix, which applies an *L*-level RTBWT decomposition. We recall that a sequence $\{\phi_n\}$ is a frame if there exist two constants $\beta \ge \alpha > 0$ such that for all **y**

$$\alpha \|\mathbf{y}\|^2 \le \sum_n |\langle \phi_n, \mathbf{y} \rangle|^2 \le \beta \|\mathbf{y}\|^2.$$
(12)

The à trous implementation of the RTBWT will enable us to show that the rows of Φ constitute a frame $\{\phi_n\}$ in \mathbb{R}^N , with the same frame bounds as the common redundant wavelet transform [18].

Proposition 1: Let $\mathbf{c} = \Phi \mathbf{y} = [\mathbf{a}_L^T, \mathbf{d}_L^T, \dots, \mathbf{d}_1^T]^T$ denote an *L*-level RTBWT decomposition of a signal \mathbf{y} . Then \mathbf{c} is a frame expansion with frame bounds $\alpha = 2$ and $\beta = 2^L$, and these bounds are the tightest possible.

The proof is presented in Appendix A.

Also, let Ψ denote an $N \times (L+1)N$ matrix that applies the *L*-level RTBWT reconstruction. Φ and Ψ apply the RTBWT analysis and synthesis operations, respectively, and $\Psi \Phi = \mathbf{I}$ where \mathbf{I} is the identity matrix. Thus Ψ is the pseudo inverse of Φ , and the columns of Ψ constitute a frame { ψ_n } which is dual to { ϕ_n }, i.e. it satisfies

$$\frac{1}{2^{L}} \|\mathbf{y}\|^{2} \leq \sum_{n} |\langle \psi_{n}, \mathbf{y} \rangle|^{2} \leq \frac{1}{2} \|\mathbf{y}\|^{2}.$$
 (13)

III. RTBWT SUBIMAGE AVERAGING FRAMES

Once the RTBWT is calculated for an image, we can apply this transform to that image, and then process it in the transform domain. However, we observed in [1] that improved results are obtained when a subimage averaging scheme is used instead. This scheme consists of applying the RTBWT to different subimages of the treated image and processing each subimage in the transform domain. Then the reconstructed image is obtained by plugging each processed subimage into its original place in the image canvas, and averaging the different values obtained for each pixel. As we show next, we can describe the transform coefficients of the different subimages as a single extended representation of the image. We construct matrices that act as analysis and synthesis operators, which are used to obtain this extended representation and reconstruct the image from it. We also show that the rows and columns of the analysis and synthesis matrices, respectively, constitute a frame and its dual.

Let $N_p = (N_1 - \sqrt{n} + 1)(N_2 - \sqrt{n} + 1)$ denote the number of overlapped patches in the image **Y**, and let **X** be an $n \times N_p$ matrix, containing column stacked versions of these patches. We extract these patches column by column, starting from the top left-most patch. When the RTBWT was constructed in the previous section, it was assumed that each patch is associated only with its middle pixel. Therefore the transform was designed to efficiently represent the signal composed of the middle points in the patches, that reside in the middle row of **X**. However, we can alternatively choose to associate all the patches with a pixel located in a different position, e.g., the top left pixel in each patch. This means that the transform can be used to represent any one of the signals located in the rows of **X**. These signals are the column stacked versions of all the *n* subimages of size $(N_1 - \sqrt{n+1}) \times (N_2 - \sqrt{n+1})$ contained in the image **Y**. We denote these subimages by $\tilde{\mathbf{Y}}_j$, j = 1, 2, ..., n.

Let Φ denote the RTBWT transform matrix, here it is of size $(L + 1)N_p \times N_p$, applying an L-level decomposition. Also, let the vector $\tilde{\mathbf{y}}_j = \mathbf{R}_j \mathbf{y}$ of length N_p be the column stacked version of $\tilde{\mathbf{Y}}_j$, where the $N_p \times N$ matrix \mathbf{R}_j extracts the *j*th subimage from the image \mathbf{y} . We first apply the RTBWT to each of the *n* subimages $\tilde{\mathbf{y}}_j$ and obtain the *n* vectors

$$\mathbf{c}_j = \Phi \tilde{\mathbf{y}}_j = \Phi \mathbf{R}_j \mathbf{y}. \tag{14}$$

We obtain the extended representation vector \mathbf{c}^{SA} by concatenating all these vectors into a single column

$$\mathbf{c}^{SA} = \begin{bmatrix} \mathbf{c}_1^T, ..., \mathbf{c}_n^T \end{bmatrix}^T = \begin{bmatrix} \Phi \mathbf{R}_1 \\ \vdots \\ \Phi \mathbf{R}_n \end{bmatrix} \mathbf{y} = \Phi^{SA} \mathbf{y} \qquad (15)$$

where we defined the matrix

$$\Phi^{SA} = \begin{bmatrix} \Phi \mathbf{R}_1 \\ \vdots \\ \Phi \mathbf{R}_n \end{bmatrix}.$$
(16)

This matrix applies the analysis operator used to obtain the representation \mathbf{c}^{SA} .

The image y is reconstructed from the representation \mathbf{c}^{SA} in the following manner. Let Ψ denote the $(L + 1)N_p \times N_p$ matrix that applies the *L*-level RTBWT reconstruction. We apply the matrix Ψ to each of the vectors \mathbf{c}_j , plug each of the obtained subimages into its original place in the image, and average the different values obtained for each pixel. More formally,

$$\mathbf{y} = \mathbf{D}^{-1} \sum_{j=1}^{n} \mathbf{R}_{j}^{T} \Psi \mathbf{c}_{j}$$
$$= \mathbf{D}^{-1} \left[\mathbf{R}_{1}^{T} \Psi, \dots, \mathbf{R}_{n}^{T} \Psi \right] \mathbf{c} = \Psi^{SA} \mathbf{c}, \qquad (17)$$

where the matrix \mathbf{R}_{j}^{T} returns the *j*th subimage into its original place in the image, and $\mathbf{D} = \sum_{j=1}^{n} \mathbf{R}_{j}^{T} \mathbf{R}_{j}$ is a diagonal weight matrix. The matrix

$$\Psi^{SA} = \mathbf{D}^{-1} \left[\mathbf{R}_1^T \Psi, \dots, \mathbf{R}_n^T \Psi \right]$$
(18)

applies the synthesis operator used to reconstruct **y** from the representation \mathbf{c}^{SA} . As we see next, the rows of Φ^{SA} constitute a frame $\{\phi_n^{SA}\}$ and the columns of Ψ^{SA} constitute its dual $\{\psi_n^{SA}\}$.

Proposition 2: The extended representation \mathbf{c}^{SA} is a frame expansion with frame bounds $\alpha = 2$ and $\beta = 2^{L}n$. The proof is presented in Appendix B.

Here we also see that Φ^{SA} and Ψ^{SA} apply analysis and synthesis operations, and $\Psi^{SA}\Phi^{SA} = \mathbf{I}$. Therefore, Ψ^{SA} is the pseudo-inverse of Φ^{SA} , and the columns of Ψ^{SA} constitute a frame { ψ_n^{SA} } which is dual to { ϕ_n^{SA} }, i.e. it satisfies

$$\frac{1}{2^L n} \|\mathbf{y}\|^2 \le \sum_n |\langle \psi_n, \mathbf{y} \rangle|^2 \le \frac{1}{2} \|\mathbf{y}\|^2.$$
(19)

Before concluding this section, we add a remark regarding the frame bounds calculated above. Let us assume that the RTBWT and corresponding frame have been calculated from an image y_0 . Then the transform is adaptive to the content in this image, since it is designed by ordering of its patches or their descendants (i.e. filtered patches), and so does the corresponding frame. Therefore, the frame bounds obtained above are in fact misleading, since inequalities such as (19) are true for a general image y, but the frame was designed to handle the image y_0 . Thus, perhaps a better definition of the frame bounds would be one that has to hold true for all the images y in a small sphere $\|\mathbf{y} - \mathbf{y}_0\|_2 \le \epsilon$, for which the transform (and the resulting frame) retain its topology. In such a case, it is quite likely that the actual frame obtained is nearly tight. We leave this matter open at this stage, as our next part of the paper does not rely directly on these frame properties.

IV. IMAGE RECONSTRUCTION USING RTBWT SA FRAMES

A. Image Reconstruction Scheme

Let **Y** be an image of size $N_1 \times N_2$ where $N_1N_2 = N$, and let **Z** be its corrupted version. Also, let **z** and **y** be the column stacked versions of **Z** and **Y**, respectively. Then we assume that the corrupted image is obtained via

$$\mathbf{z} = \mathbf{H}\mathbf{y} + \mathbf{v},\tag{20}$$

where the $N \times N$ matrix **H** denotes a linear operator that corrupts the data, and **v** denotes an additive white Gaussian noise independent of **y** with zero mean and variance σ^2 . This setting can be used to describe several of the classic image inverse problems such as image denoising, deblurring, and inpainting. Our goal is to reconstruct **y** from **z** by optimizing an objective function that uses the analysis and synthesis operators¹ Φ^{SA} and Ψ^{SA} as sparsity promoting regularizers.

A common approach for image reconstruction is to solve an inverse problem which uses either analysis-based or synthesisbased image priors [19]. Here we take a different approach, follow the footsteps of [11], and consider the following combination of an analysis and a synthesis problem

$$\{\hat{\mathbf{y}}, \hat{\mathbf{c}}\} = \underset{\mathbf{y}, \mathbf{c}}{\operatorname{argmin}} \|\mathbf{z} - \mathbf{H}\mathbf{y}\|_{2}^{2} + \eta \|\mathbf{y} - \Psi\mathbf{c}\|_{2}^{2} + \lambda \|\mathbf{c}\|_{0} + \mu \|\mathbf{c} - \Phi\mathbf{y}\|_{2}^{2}.$$
(21)

Indeed, it can be seen that the problem (21) reduces to a synthesis problem when $\eta \to \infty$ and $\mu = 0$, and to an analysis when $\eta = 0$ and $\mu \to \infty$. The authors of [11] brilliantly handle the problem (21) as a generalized Nash equilibrium (GNE) process, and split it into two subproblems, a denoising task:

$$\hat{\mathbf{c}} = \underset{\mathbf{c}}{\operatorname{argmin}} \ \frac{\lambda}{\mu} \|\mathbf{c}\|_0 + \|\mathbf{c} - \Phi \mathbf{y}\|_2^2, \tag{22}$$

and an inversion task:

$$\hat{\mathbf{y}} = \underset{\mathbf{y}}{\operatorname{argmin}} \|\mathbf{z} - \mathbf{H}\mathbf{y}\|_{2}^{2} + \eta \|\mathbf{y} - \Psi\mathbf{c}\|_{2}^{2}$$
(23)

¹As of this point in the paper, we will simplify our notations and use Φ , Ψ and **c**, referring to the full adaptive frames Φ^{SA} and Ψ^{SA} and the representation \mathbf{c}^{SA} .

Algorithm 4 Image Reconstruction Scheme

Task: Recover the image \mathbf{y} from the corrupted image \mathbf{z} . **Parameters:** The corrupted image \mathbf{z} , an initial estimate \mathbf{y}_{init} , the matrix \mathbf{H} , the iteration number G, the threshold τ , and the design parameter η . **Initialization:** Set $\mathbf{y}_0 = \mathbf{y}_{init}$ and use it to calculate Φ and Ψ .

Main Iteration: For t = 1, ..., G perform:

Denoising:
$$\mathbf{c}_t = S_{\tau} \{ \mathbf{\Phi} \mathbf{y}_{t-1} \}$$

• Inversion: $\mathbf{y}_t = [\mathbf{H}^T \mathbf{H} + \eta \mathbf{I}]^{-1} [\mathbf{H}^T \mathbf{z} + \eta \Psi \mathbf{c}_t]$

Output: The estimate $\hat{\mathbf{y}} = \mathbf{y}_G$.

which are solved sequentially. We note that Φ and Ψ are calculated in the beginning of this iterative process, and one approach is to keep them fixed throughout the iterative process. Alternatively, one might update these operators based on the updated image. Unless said otherwise, our scheme takes the first path of fixing these frames during the iterations.

The solution of the inversion stage is simply

$$\hat{\mathbf{y}} = \left[\mathbf{H}^T \mathbf{H} + \eta \mathbf{I}\right]^{-1} \times \left[\mathbf{H}^T \mathbf{z} + \eta \Psi \mathbf{c}\right].$$
(24)

The solution of the denoising problem is also quite simple, given by

$$\hat{\mathbf{c}} = S^{hard} \left\{ \Phi \mathbf{y} \right\} \tag{25}$$

where the operator S^{hard} applies hard thresholding with the threshold $\sqrt{\lambda/\mu}$. We choose to replace S^{hard} with an operator S_{τ} that sets to zero coefficients in **c** in such a manner that the thresholded versions of the coefficient vectors **c**_j (see Equation (14)) share a common support. This way, the set of thresholded subimage coefficients form a joint sparsity pattern in the transform domain, and our experiments show that when such patterns are used, the quality of the reconstructed images improves. Let $c_j[k]$ be the *k*th sample in the coefficient vector **c**_j. Then the operator S_{τ} goes over the indices k = $1, \ldots, (L + 1)N_p$, and for each index sets the coefficients $\{c_1[k], \ldots, c_n[k]\}$ to zero if

$$\sqrt{\frac{1}{n}\sum_{j=1}^{n}c_{j}^{2}[k]} < \tau\sigma \tag{26}$$

where τ is a design parameter. This corresponds to a simple modification in the original inverse problem formulation in Equation (21), replacing the term $\|\mathbf{c}\|_0$ by a mixed norm applied on the different pieces of this representation, namely, $\|\mathbf{c}\|_{2,0}$.

The obtained image reconstruction scheme is described in Algorithm 4. It can be seen that this algorithm is similar to the IDD-BM3D algorithm proposed in [11], to which we will later compare our image deblurring results. We next demonstrate the use of the scheme in Algorithm 4 for image denoising and deblurring. In all the experiments described next we choose the distance function w to be the Euclidean distance, and use a 9-level RTBWT decomposition with the Symmlet 8 wavelet filter.

B. Image Denoising

In the case of image denoising $\mathbf{z} = \mathbf{y} + \mathbf{v}$ and therefore $\mathbf{H} = \mathbf{I}$. We perform denoising using a simplified version of the scheme in Algorithm 4, which uses the initial estimate $\mathbf{y}_{init} = \mathbf{z}$, and applies only one iteration. Also, we notice that since $\mathbf{H} = \mathbf{I}$, the solution to the inversion problem (23) becomes:

$$\hat{\mathbf{y}} = \frac{1}{1+\eta} \mathbf{z} + \frac{\eta}{1+\eta} \Psi \mathbf{c}.$$
 (27)

which is a weighted average between a denoised image and the original noisy image. We further simplify our scheme and cancel the addition of the noisy data to the clean image by setting $\eta \rightarrow \infty$. All these simplifications were done since we found experimentally that the resulting simplified algorithm is already producing near state-of-the-art results. Thus, the obtained denoising scheme consists of calculating $\mathbf{c} = \Phi \mathbf{z}$, applying to it the thresholding operator S_{τ} , and reconstructing the image using Ψ , i.e.

$$\hat{\mathbf{y}} = \Psi S_{\tau} \{ \mathbf{c} \} = \Psi S_{\tau} \{ \Phi \mathbf{z} \}.$$
(28)

We note that aside from the use of the operator S_{τ} , this scheme is similar to the common thresholding based denoising procedure for signal and image processing using wavelet frames.

We explore two different methods to further improve the results obtained with the denoising algorithm described above. Both these methods apply the above scheme, and then use the patches from the "cleaned" result $\hat{\mathbf{y}}^1$ to construct a "better version" of the RTBWT, and use this transform to calculate new analysis and synthesis operators Φ^1 and Ψ^1 . The first method obtains the denoised image $\hat{\mathbf{y}}^2$ by applying Equation (28) again with these modified operators, i.e.

$$\hat{\mathbf{y}}^2 = \Psi^1 S_\tau \{ \Phi^1 \mathbf{z} \}. \tag{29}$$

The second method applies a different scheme in its second stage. Let \mathbf{C}_{diag} be a diagonal matrix that contains the squares of the elements of the vector $\Phi^1 \hat{\mathbf{y}}^1$ in its main diagonal. Then we replace Φ with Φ^1 and the ℓ_0 -norm term $\|\mathbf{c}\|_0$ with a different sparsity promoting term $\mathbf{c}^T \mathbf{C}_{diag}^{-1} \mathbf{c}$ in the denoising problem (22). We obtain the new problem

$$\hat{\mathbf{c}} = \underset{\mathbf{c}}{\operatorname{argmin}} \ \frac{\lambda}{\mu} \mathbf{c}^T \mathbf{C}_{diag}^{-1} \mathbf{c} + \|\mathbf{c} - \Phi^1 \mathbf{y}\|_2^2.$$
(30)

We differentiate (30) with respect to \mathbf{c} , equate the derivative the zero, and obtain

$$\hat{\mathbf{c}} = \left[\frac{\lambda}{\mu}\mathbf{C}_{diag}^{-1} + \mathbf{I}\right]^{-1}\Phi^{1}\mathbf{y}$$

$$= \left[\frac{\lambda}{\mu}\mathbf{C}_{diag}^{-1} + \mathbf{C}_{diag}^{-1}\mathbf{C}_{diag}\right]^{-1}\Phi^{1}\mathbf{y}$$

$$= \left[\mathbf{C}_{diag} + \frac{\lambda}{\mu}\mathbf{I}\right]^{-1}\mathbf{C}_{diag}\Phi^{1}\mathbf{y} = \mathbf{W}\Phi^{1}\mathbf{y}, \quad (31)$$

where we have defined the diagonal matrix

$$\mathbf{W} = \left[\mathbf{C}_{diag} + \frac{\lambda}{\mu}\mathbf{I}\right]^{-1}\mathbf{C}_{diag}.$$
 (32)



Fig. 7. (a) PSNR gain versus the number of scales in which permutations are applied. (b) PSNR values obtained for different sizes of B for noisy versions of the images Lena and Barbara (Barb) with noise standard deviations of 10, 25 and 50 (in brackets).

TABLE I Parameters Used in the Denoising Experiments

| σ | Stage | L | \sqrt{n} | B | au |
|----------|--------|---|------------|-----|------|
| | 1 | 9 | 7 | 21 | 1.5 |
| 5 | 2 | 9 | 4 | 211 | 1.45 |
| | Wiener | 9 | 5 | 191 | 1.05 |
| | 1 | 9 | 9 | 21 | 1.5 |
| 10 | 2 | 9 | 6 | 181 | 1.45 |
| | Wiener | 9 | 5 | 191 | 1.05 |
| | 1 | 9 | 10 | 21 | 1.5 |
| 15 | 2 | 9 | 7 | 211 | 1.45 |
| | Wiener | 9 | 5 | 191 | 1.05 |
| | 1 | 9 | 11 | 21 | 1.5 |
| 20 | 2 | 9 | 9 | 211 | 1.45 |
| | Wiener | 9 | 5 | 191 | 1.05 |
| 25 | 1 | 9 | 12 | 21 | 1.5 |
| | 2 | 9 | 10 | 211 | 1.45 |
| | Wiener | 9 | 6 | 191 | 1.05 |
| | 1 | 9 | 14 | 21 | 1.5 |
| 50 | 2 | 9 | 13 | 211 | 1.45 |
| | Wiener | 9 | 7 | 191 | 1.05 |
| | 1 | 9 | 16 | 21 | 1.5 |
| 75 | 2 | 9 | 14 | 211 | 1.45 |
| | Wiener | 9 | 7 | 191 | 1.05 |
| | 1 | 9 | 16 | 21 | 1.5 |
| 100 | 2 | 9 | 16 | 211 | 1.45 |
| | Wiener | 9 | 9 | 191 | 1.05 |

Thus, the second method obtains the denoised image $\hat{\mathbf{y}}_w^2$ by calculating in its second stage the coefficient vector $\mathbf{c} = \Phi^1 \mathbf{z}$, multiplying it with the matrix \mathbf{W} , and reconstructing the image using Ψ^1 , i.e.

$$\hat{\mathbf{y}}_{w}^{2} = \Psi^{1} \mathbf{W} \Phi^{1} \mathbf{z}. \tag{33}$$

We note that by choosing $\lambda/\mu = (\tau \sigma)^2$, where τ is a design parameter, multiplication by the matrix **W** is equivalent to applying some sort of a Wiener filter to the noisy image, similarly to what is done in the second stage of the algorithms in [9] and [20].

We now proceed to assess the performance of our image denoising scheme. We note that in all our experiments the noise levels are assumed to be known. We start by demonstrating the performance gain obtained by applying the permutations in all the scales of the RTBWT. We apply the following procedure to the noisy versions of the images Lena and Barabra with noise standard deviation $\sigma = 25$, that are shown in the first column of Figure 8. For each image we calculate 10 different variants of the RTBWT with the parameters $\sqrt{n} = 12$ and B = 21: one with no permutations at



Fig. 8. Denoising results (PSNR) for the images Lena and Barbara ($\sigma = 25$, input PSNR = 20.18 dB): First column: noisy images, Second column - BM3D results, Third column - 2 stages results of the RTBWT denoising scheme, Fourth column - results of the RTBWT scheme followed by Wiener filtering.

DENOISING RESULTS (PSNR IN dB) OF NOISY VERSIONS OF 6 IMAGES, OBTAINED WITH THE BM3D ALGORITHM, TWO STAGES OF THE RTBWT DENOISING SCHEME, AND THE RTBWT SCHEME FOLLOWED BY WIENER FILTERING. FOR EACH IMAGE AND NOISE LEVEL THE BEST RESULT AND RESULTS WITHIN A DISTANCE OF 0.05 dB FROM IT, ARE HIGHLIGHTED

TABLE II

| Terrore | Mada a | σ/PSNR | | | | | | | |
|-------------|---------------------------|---------|----------|----------|----------|----------|----------|----------|----------|
| mage | Method | 5/34.16 | 10/28.14 | 15/24.61 | 20/22.11 | 25/20.18 | 50/14.16 | 75/10.63 | 100/8.14 |
| | BM3D | 38.72 | 35.93 | 34.27 | 33.05 | 32.08 | 29.05 | 27.26 | 25.95 |
| Lana | proposed (1 stage) | 38.44 | 35.70 | 34.12 | 32.98 | 32.06 | 28.97 | 27.19 | 25.90 |
| Lena | proposed (2 stages) | 38.08 | 35.45 | 33.91 | 32.99 | 32.11 | 29.18 | 27.43 | 26.31 |
| | proposed (1 stage+Wiener) | 38.40 | 35.75 | 34.22 | 33.11 | 32.26 | 29.30 | 27.50 | 26.36 |
| | BM3D | 38.31 | 34.98 | 33.11 | 31.78 | 30.72 | 27.23 | 25.12 | 23.62 |
| Doubouo | proposed (1 stage) | 37.54 | 34.50 | 32.89 | 31.70 | 30.73 | 27.39 | 25.38 | 23.97 |
| Darbara | proposed (2 stages) | 37.62 | 34.55 | 32.91 | 31.71 | 30.76 | 27.65 | 25.73 | 24.44 |
| | proposed (1 stage+Wiener) | 37.67 | 34.55 | 32.97 | 31.85 | 30.90 | 27.78 | 25.82 | 24.46 |
| | BM3D | 37.28 | 33.92 | 32.14 | 30.88 | 29.91 | 26.78 | 25.12 | 23.97 |
| Poets | proposed (1 stage) | 36.86 | 33.68 | 31.91 | 30.65 | 29.67 | 26.59 | 24.85 | 23.68 |
| Boats | proposed (2 stages) | 36.86 | 33.64 | 31.92 | 30.67 | 29.71 | 26.80 | 25.13 | 24.03 |
| | proposed (1 stage+Wiener) | 36.98 | 33.79 | 32.05 | 30.82 | 29.88 | 26.91 | 25.15 | 24.04 |
| Fingerprint | BM3D | 36.51 | 32.46 | 30.28 | 28.81 | 27.70 | 24.53 | 22.83 | 21.61 |
| | proposed (1 stage) | 35.03 | 31.28 | 29.47 | 28.21 | 27.16 | 23.90 | 22.20 | 21.19 |
| | proposed (2 stages) | 35.86 | 31.88 | 29.80 | 28.17 | 27.01 | 23.74 | 22.22 | 21.37 |
| | proposed (1 stage+Wiener) | 35.38 | 31.60 | 29.73 | 28.45 | 27.32 | 24.06 | 22.47 | 21.53 |
| House | BM3D | 39.83 | 36.71 | 34.94 | 33.77 | 32.86 | 29.69 | 27.51 | 25.87 |
| | proposed (1 stage) | 39.32 | 36.41 | 34.75 | 33.61 | 32.74 | 29.55 | 27.43 | 25.75 |
| | proposed (2 stages) | 38.40 | 35.68 | 34.01 | 33.27 | 32.46 | 29.49 | 27.42 | 25.96 |
| | proposed (1 stage+Wiener) | 38.42 | 35.86 | 34.25 | 33.08 | 32.37 | 29.56 | 27.37 | 25.98 |
| Peppers | BM3D | 38.12 | 34.68 | 32.70 | 31.29 | 30.16 | 26.68 | 24.73 | 23.39 |
| | proposed (1 stage) | 37.65 | 34.27 | 32.36 | 30.99 | 29.90 | 26.48 | 24.50 | 23.08 |
| | proposed (2 stages) | 37.64 | 34.44 | 32.61 | 31.21 | 30.14 | 26.76 | 24.89 | 23.53 |
| | proposed (1 stage+Wiener) | 37.81 | 34.49 | 32.66 | 31.35 | 30.33 | 26.93 | 24.98 | 23.56 |

all and the rest with all the permutations in the first *F* scales, $1 \le F \le 9$. We then apply to each image one stage of the aforementioned denoising scheme with each of the 10 different variants of the RTBWT and with the parameter $\tau = 1.5$. Figure 7(a) shows the PSNR values obtained for each image versus the number of scales in which permutations are applied. It can be seen that most of the performance gain is obtained by the permutation in the first scale ($\ell = 0$), but the PSNR increases as more permutations are used in coarser scales as well.

We next demonstrate the robustness of our denoising performance with respect to the size of the parameter B. We apply to noisy versions of the images Lena and Barbara, with noise standard deviations of 10, 25 and 50, one stage of our denoising scheme with \sqrt{n} sizes of 9,12 and 14, respectively, and varying values of B. Figure 7(b) shows 6 curves, one for each combination of image and noise level, describing the PSNR values obtained for different sizes of the parameter B. First it can be seen that all the curves obtain their maximum values for sizes of B between 10 to 50, and that the change in the PSNR in that range is relatively moderate. Thus high quality denoising results may be obtained by choosing a single value of B in that range, and applying our denoising scheme with it to all the noisy images with the different noise levels. It can also be seen that after the curves pass their maxima, their PSNR values decrease as the size of B increases. However, even in the case where B equals the size of the whole image our denoising scheme manages to obtain a high quality denoising result.

Now, in order to assess the performance of the proposed image denoising scheme we apply it to noisy versions of 6 images, with 8 different noise standard deviations. The parameters employed by the proposed denoising scheme for the different noise levels are shown in Table I. Table II shows the PSNR values of the results obtained with the BM3D algorithm, two stages of the RTBWT denoising scheme, and the RTBWT followed by Wiener filtering. The noisy and recovered images obtained for $\sigma = 25$ with the BM3D algorithm, and our results are shown in Figure 8. First, it can be seen that both the second stage of the RTBWT scheme and the Wiener filter improve the first iteration results of the RTBWT scheme in many (70% - 85%) of the cases. The Wiener filter results are generally better or comparable to the results of both the first and second stages of the RTBWT scheme. Finally it can be seen that for $\sigma \ge 25$, the RTBWT denoising scheme followed by Wiener filtering achieves in most of the cases either the best results or results that are nearly as good. As the noise decreases in strength, the performance of the RTBWT based denoising schemes deteriorates compared to the BM3D.

In terms of visual quality both the BM3D and the RTBWTbased algorithms produce high quality results, however the two approaches also suffer from different types of artifacts: the BM3D results suffer from smudges (for example in Lena's face) while the RTBWT-based algorithms suffer from false contours (for example around Lena's hat).

C. Image Deblurring

In the case of image deblurring, H is a blur matrix, and we perform deblurring by simply applying Algorithm 4 as is. We demonstrate the image deblurring performance obtained with this algorithm on the images Lena, Barbara, House, and Cameraman, for the 6 scenarios described in Table III. We compare the results obtained with this algorithm to the ones obtained with the BM3DDEB [10] and the IDD-BM3D algorithms. As the IDD-BM3D algorithm is initialized with the BM3DDEB results, for a fair comparison we examine the results obtained with our scheme using the same initialization. We also examine the results obtained with our scheme by initializing it with the blurry image, and try to improve them by applying two more stages of our scheme, where we initialize each stage with result of the previous ones. The parameters employed in the different stages of the proposed deblurring scheme, with the different initializations and for the different scenarios, are shown in Table IV. We note that similarly to what was done in [11], we optimized the parameters η and τ separately for each stage and each deblurring scenario to provide best reconstruction quality.

TABLE III Blur Point Spread Functions (PSF) and Noise Variances Used in the Different Deblurring Scenarios

| Scenario | PSF | σ^2 |
|----------|---|---------------|
| 1 | $1/(1+x_1^2+x_2^2)$, $x_1, x_2 = -7, \dots, 7$ | 2 |
| 2 | $1/(1+x_1^2+x_2^2)$, $x_1, x_2 = -7, \dots, 7$ | 8 |
| 3 | 9×9 uniform | ≈ 0.3 |
| 4 | $[1 4 6 4 1]^T [1 4 6 4 1] / 256$ | 49 |
| 5 | Gaussian with $std = 1.6$ | 4 |
| 6 | Gaussian with $std = 0.4$ | 64 |

TABLE IV

PARAMETERS USED IN THE DEBLURRING EXPERIMENTS FOR THE PROPOSED SCHEME INITIALIZED WITH THE BM3DDEB RESULTS (BM3DDEB INIT.) AND FOR 3 STAGES OF THE PROPOSED SCHEME INITIALIZED WITH THE BLURRY IMAGE (BI)

| Scenario | Stage | $\mid L$ | \sqrt{n} | B | η | $ \tau$ | G |
|----------|-------------------|----------|------------|----|--------|----------|----|
| 1 | 1 (BM3DDEB init.) | 9 | 3 | 51 | 0.07 | 20 | 50 |
| | 1 (BI) | 9 | 6 | 31 | 0.03 | 60 | 30 |
| 1 | 2(BI) | 9 | 4 | 41 | 0.03 | 50 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 0.06 | 30 | 30 |
| | 1 (BM3DDEB init.) | 9 | 3 | 51 | 0.08 | 30 | 50 |
| 2 | 1 (BI) | 9 | 6 | 31 | 0.07 | 60 | 30 |
| 2 | 2 (BI) | 9 | 4 | 41 | 0.1 | 50 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 0.16 | 30 | 30 |
| | 1 (BM3DDEB init.) | 9 | 3 | 51 | 0.01 | 20 | 50 |
| 2 | 1 (BI) | 9 | 6 | 31 | 0.005 | 50 | 30 |
| 5 | 2 (BI) | 9 | 4 | 41 | 0.005 | 60 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 0.015 | 30 | 30 |
| | 1 (BM3DDEB init.) | 9 | 3 | 51 | 0.9 | 20 | 50 |
| 4 | 1 (BI) | 9 | 6 | 31 | 0.5 | 50 | 30 |
| | 2 (BI) | 9 | 4 | 41 | 0.8 | 40 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 1.9 | 20 | 30 |
| | 1 (BM3DDEB init.) | 9 | 3 | 51 | 0.05 | 20 | 50 |
| 5 | 1 (BI) | 9 | 6 | 31 | 0.03 | 60 | 30 |
| 5 | 2 (BI) | 9 | 4 | 41 | 0.07 | 40 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 0.14 | 20 | 30 |
| 6 | 1 (BM3DDEB init.) | 9 | 3 | 51 | 3.7 | 20 | 50 |
| | 1 (BI) | 9 | 6 | 31 | 3.2 | 50 | 30 |
| | 2 (BI) | 9 | 4 | 41 | 3.9 | 30 | 30 |
| | 3 (BI) | 9 | 4 | 41 | 3.8 | 30 | 30 |

Table V shows the ISNR results obtained with the different algorithms. The blurred and recovered Lena and Cameraman images obtained with the BM3DDEB and IDD-BM3D algorithms, our proposed scheme initialized with the BM3DDEB results, and three stages of the proposed scheme initialized with the blurry images, are shown in Figure 9. It can be seen that for all the scenarios except for scenario 3, the proposed scheme initialized with BM3DDEB results achieves the best performance for every image. Our scheme obtains inferior results when the blurry image is used to initialize it, however its performance improve when two more rounds are applied. In fact, in some cases three rounds of our scheme initialized with the blurry image obtain the best or second best results.

In terms of visual quality, again both the BM3D-based and the RTBWT-based algorithms produce high quality results, accompanied by modest and different types of artifacts: here the results of the BM3D-based algorithms suffer from false contouring (for example near the Cameraman's chin) while those of the RTBWT-based algorithms suffer from a more significant noise residual.



Fig. 9. Deblurring results (ISNR) for the images Lena and Cameraman (Scenario 2): First column: blurry images, Second column: BM3DDEB results, Third column: IDD-BM3D results, Fourth column: results of the proposed scheme initialized with the BM3DDEB results, Fifth column: results of the three rounds of the proposed scheme initialized with the blurry images.

TABLE V

DEBLURRING RESULTS (ISNR IN dB) OF BLURRY VERSIONS OF 4 IMAGES, OBTAINED WITH THE BM3DDEB AND IDD-BM3D ALGORITHMS, THE PROPOSED SCHEME INITIALIZED WITH THE BM3DDEB RESULTS, AND 3 STAGES OF THE PROPOSED SCHEME INITIALIZED WITH THE BLURRY IMAGE (BI). FOR EACH IMAGE AND SCENARIO THE BEST RESULT AND RESULTS WITHIN A DISTANCE OF 0.05 dB FROM IT, ARE HIGHLIGHTED

| Image | Method | Scenario 1 | Scenario 2 | Scenario 3 | Scenario 4 | Scenario 5 | Scenario 6 |
|-----------|--------------------------|------------|------------|------------|------------|------------|------------|
| | BM3DDEB | 7.96 | 6.55 | 7.99 | 4.80 | 4.34 | 6.42 |
| | IDD-BM3D | 7.97 | 6.61 | 8.91 | 4.97 | 4.85 | 6.34 |
| Lena | proposed (BM3DDEB init.) | 8.56 | 6.92 | 8.86 | 5.52 | 4.95 | 6.91 |
| | proposed (BI 1 iter.) | 7.36 | 5.99 | 8.43 | 4.66 | 4.56 | 5.58 |
| | proposed (BI 2 iter.) | 7.43 | 6.25 | 8.69 | 4.76 | 4.83 | 5.89 |
| | proposed (BI 3 iter.) | 7.75 | 6.34 | 8.98 | 4.88 | 4.91 | 5.89 |
| | BM3DDEB | 7.88 | 4.13 | 5.91 | 2.05 | 1.29 | 5.85 |
| | IDD-BM3D | 7.64 | 3.96 | 6.05 | 1.88 | 1.16 | 5.45 |
| Barbara | proposed (BM3DDEB init.) | 8.06 | 4.57 | 6.01 | 2.20 | 1.41 | 6.06 |
| | proposed (BI 1 iter.) | 4.98 | 2.54 | 4.52 | 1.48 | 0.97 | 4.59 |
| | proposed (BI 2 iter.) | 5.98 | 2.68 | 4.76 | 1.48 | 0.99 | 4.88 |
| | proposed (BI 3 iter.) | 6.31 | 2.76 | 4.99 | 1.49 | 1.00 | 4.88 |
| | BM3DDEB | 9.34 | 8.22 | 10.94 | 5.20 | 4.59 | 7.34 |
| | IDD-BM3D | 9.95 | 8.55 | 12.89 | 5.79 | 5.74 | 7.13 |
| House | proposed (BM3DDEB init.) | 10.44 | 8.79 | 13.11 | 6.38 | 5.95 | 7.56 |
| | proposed (BI 1 iter.) | 9.09 | 7.58 | 12.03 | 5.27 | 5.30 | 6.16 |
| | proposed (BI 2 iter.) | 9.10 | 7.81 | 12.69 | 5.29 | 5.58 | 6.52 |
| | proposed (BI 3 iter.) | 9.58 | 8.01 | 13.16 | 5.58 | 5.74 | 6.56 |
| | BM3DDEB | 8.20 | 6.46 | 8.37 | 3.35 | 3.72 | 4.70 |
| | IDD-BM3D | 8.85 | 7.12 | 10.45 | 3.98 | 4.31 | 4.89 |
| Cameraman | proposed (BM3DDEB init.) | 9.24 | 7.38 | 10.21 | 4.34 | 4.68 | 5.26 |
| | proposed (BI 1 iter.) | 7.92 | 6.18 | 8.19 | 3.46 | 3.97 | 3.92 |
| | proposed (BI 2 iter.) | 8.27 | 6.65 | 9.57 | 3.80 | 4.39 | 4.27 |
| | proposed (BL 3 iter.) | 8.29 | 6.69 | 9.81 | 3.94 | 4.58 | 4.26 |

D. Computational Complexity

We next evaluate the computational complexity of the image denoising and deblurring algorithms described above. We start by calculating the complexity of a single iteration of the denoising algorithm. First, extracting all the overlapped image patches requires O(nN) operations. Next, we assume that for the calculation of each one of the 2^{ℓ} operators \mathbf{P}_{ℓ}^{s} in the ℓ th scale of the RTBWT, each patch requires approximately B^{2} distance calculations. As calculating the Euclidean distance between two patches requires O(n) operations, the number of operations required to calculate a single reordering operator \mathbf{P}_{ℓ}^{s} from $\tilde{N}_{\ell} = N2^{-\ell}$ patches is approximately $O(\tilde{N}_{\ell}B^{2}n)$. Reordering these patches and applying them the scaling and wavelet filters requires $O(n\tilde{N}_{\ell})$ operations. Therefore the number of operations required in order to calculate from an image all the operators \mathbf{P}_{ℓ}^{s} used in an *L*-level RTBWT is

$$O(nN) + \sum_{\ell=0}^{L-1} 2^{\ell} \left[O(\tilde{N}_{\ell} B^2 n) + O(n\tilde{N}_{\ell}) \right] = O(LNB^2 n).$$
(34)

Now, reordering each signal \mathbf{a}_{ℓ}^{s} and applying on it the scaling and wavelet filters requires $O(\tilde{N}_{\ell})$ operations, therefore

applying either of the analysis and synthesis operators Φ and Ψ , corresponding to the RTBWT calculated above, requires $\sum_{\ell=0}^{L-1} 2^{\ell} O(\tilde{N}_{\ell}) = O(LN)$ operations.

Next, each of the following actions require O(nLN) operations: applying the operator Φ to the *n* subimages $\tilde{\mathbf{z}}_j$, applying the threshold operator S_τ or the Wiener filter to the transform coefficients, and applying Ψ to the result, Constructing an estimate image by averaging the pixel values obtained with the different subimages requires O(nN) operations, therefore the total complexity of a single denoising iteration is

$$O(LNB^2n) + O(nLN) + O(nN) = O(LNB^2n)$$
(35)

operations, which means that, as might be expected, the overall complexity is dominated by the calculation of the RTBWT permutation operators. For a typical case in our experiments, $N = 512^2$, L = 9, n = 121 and B = 21, the above amounts to $1.26 \cdot 10^{11}$ operations.

We proceed to calculate the complexity of the proposed image deblurring algorithm. As we saw above, the calculation of all the RTBWT operators requires $O(LNB^2n)$ operations, and each denoising step requires O(nLN) + O(nN) = O(nLN) operations. The deblurring step is performed in the Fourier domain and therefore requires $O(N \log N)$ operations. Thus, the overall complexity of the algorithm is

$$O(LNB^{2}n) + G[O(nLN) + O(N \log N)]$$

= $O(LNB^{2}n + GnLN + GN \log N)$ (36)

and since the parameter we use satisfy $nLB^2 > G(nL + \log N)$, the complexity of the algorithm is again dominated by the calculation of the RTBWT permutation operators. For a typical case in our experiments, $N = 512^2$, L = 9, n = 9, B = 51 and G = 50, the above amounts to $5.64 \cdot 10^{10}$ operations.

In order to better illustrate these numbers, we also provide run-times: applying two iterations of our denoising scheme to a 512 × 512 image with noise level $\sigma = 25$ using a non optimized and non parallel matlab implementation, on an Intel(R) Core(TM) i7-2600 CPU @ 3.40 GHz, takes about 45 minutes. However, applying a single iteration followed by Wiener filtering takes only about 20 minutes. Applying a single round of our deblurring scheme, initialized with the BM3DDEB result, to a 512 × 512 image corrupted according to scenario 1 takes about 4 minutes. We should note that while in our experiments we employed exact exhaustive search, approximate nearest neighbor algorithms may be used to alleviate the computational burden.

V. CONCLUSIONS

We have revisited the redundant tree-based wavelet transform proposed in [1], described it in greater details and analyzed its properties. We have introduced an alternative implementation for this transform which is based on the à trous algorithm, and used it to show that the RTBWT is a frame. We have also shown that the image representation obtained with the RTBWT combined with the subimage averaging scheme also constitutes a frame, and calculated the synthesis and analysis operators associated with it. We have proposed image denoising and deblurring algorithms which make use of these operators as sparsity promoting regularizers, and demonstrated state-of-the-art results.

There are several research directions to extend this work that we are currently considering. The first is to explore the use of a weighted average in the subimage averaging scheme. Intuitively it seems that the RTBWT would better represent subimages which reside closer to the center of the image, and therefore their pixels should receive higher weights when the final estimate is calculated. A different direction is to replace the thresholding operator S_{τ} used in the proposed image processing schemes with an operator that takes into consideration the change in the noise levels in the different wavelet scales. Further, the operators Φ and Ψ may be used to solve different image processing problems such as image inpainting, superresolution, tomographic reconstruction, and more. Finally, it might be interesting to assess the performance gain (if any), obtained by a non-simplified version of our image denoising scheme, that uses finite values of η and more than one iteration in each stage.

APPENDIX

A. Proof of Proposition 1

Our proof is similar in spirit to the one given in [18] for the common redundant wavelet transform. By the definition of a frame, it is sufficient to show that the frame bounds exist in order to show that the RTBWT is a frame. We notice that $\|\mathbf{c}\|^2 = \sum_n |\langle \phi_n, \mathbf{y} \rangle|^2$, and we next show that the coefficient vector \mathbf{c} satisfies

$$2\|\mathbf{y}\|^{2} \le \|\mathbf{c}\|^{2} \le 2^{L} \|\mathbf{y}\|^{2}.$$
 (A.1)

We first recall that the orthogonal scaling and wavelet filters satisfy [16]

$$\bar{h}[n] = h[-n] \text{ and } \bar{g}[n] = g[-n]$$
 (A.2)

$$\sum_{n} h[n] = \sqrt{2} \text{ and } \sum_{n} g[n] = 0$$
 (A.3)

and that the filters \mathbf{h} and \mathbf{g} are power complementary, i.e. their Fourier transforms satisfy

$$|\hat{h}(2^{\ell}\omega)|^{2} + |\hat{g}(2^{\ell}\omega)|^{2} = 2.$$
 (A.4)

We apply the Fourier transform to (8) and (9), and using (A.2) we obtain that

$$\hat{a}_{\ell+1}(\omega) = \hat{a}_{\ell}^{p}(\omega)\hat{h}^{*}(2^{\ell}\omega) \tag{A.5}$$

$$\hat{d}_{\ell+1}(\omega) = \hat{a}_{\ell}^p(\omega)\hat{g}^*(2^{\ell}\omega). \tag{A.6}$$

Next we use (A.4) and the fact that $\|\mathbf{a}_{\ell}^{p}\|^{2} = \|\mathbf{a}_{\ell}\|^{2}$ and obtain that

$$\|\mathbf{d}_{\ell+1}\|^{2} + \|\mathbf{a}_{\ell+1}\|^{2} = \frac{1}{2\pi} \int_{-\pi}^{\pi} |a_{\ell}^{p}(\omega)|^{2} \left(|\hat{h}(2^{\ell}\omega)|^{2} + |\hat{g}(2^{\ell}\omega)|^{2} \right) d\omega = 2\|\mathbf{a}_{\ell}\|^{2}.$$
(A.7)

We notice that

$$\|\mathbf{c}\|^{2} = \|\mathbf{a}_{L}\|^{2} + \|\mathbf{d}_{L}\|^{2} + \sum_{\ell=1}^{L-1} \|\mathbf{d}_{\ell}\|^{2}$$
$$= 2\|\mathbf{a}_{L-1}\|^{2} + \|\mathbf{d}_{L-1}\|^{2} + \sum_{\ell=1}^{L-2} \|\mathbf{d}_{\ell}\|^{2} \qquad (A.8)$$

where we used (A.7) in the transition from the first to the second line. Finally from (A.7) and (A.8) and the fact that $\mathbf{a}_0 = \mathbf{y}$ we get that

$$\|\mathbf{c}\|^{2} \leq 2\|\mathbf{a}_{L-1}\|^{2} + 2\|\mathbf{d}_{L-1}\|^{2} + \sum_{\ell=1}^{L-2} \|\mathbf{d}_{\ell}\|^{2}$$
(A.9)
$$\leq 2^{2}\|\mathbf{a}_{L-2}\|^{2} + 2^{2}\|\mathbf{d}_{L-2}\|^{2} + \sum_{\ell=1}^{L-3} \|\mathbf{d}_{\ell}\|^{2} \leq \ldots \leq 2^{L}\|\mathbf{y}\|^{2}$$
(A.10)

and that

$$\|\mathbf{c}\|^{2} \geq \|\mathbf{a}_{L-1}\|^{2} + \|\mathbf{d}_{L-1}\|^{2} + \sum_{\ell=1}^{L-2} \|\mathbf{d}_{\ell}\|^{2}$$
(A.11)
$$\geq \|\mathbf{a}_{L-2}\|^{2} + \|\mathbf{d}_{L-2}\|^{2} + \sum_{\ell=1}^{L-3} \|\mathbf{d}_{\ell}\|^{2} \geq \ldots \geq 2\|\mathbf{y}\|^{2}.$$
(A.12)

We now show that the bounds $\alpha = 2$ and $\beta = 2^L$ are the tightest possible frame bounds since we can find vectors that meet them. We first show that the vector \mathbf{y}_N , which satisfies $y_N[k] = \frac{1}{\sqrt{N}}$, k = 1..., N meets the upper bound. Since $\mathbf{a}_0 = \mathbf{y}_N$ is a constant signal, we get that $\mathbf{a}_0^{1,p} = \mathbf{a}_0 = \mathbf{y}_N$. Using (A.2) and (A.3) we get that

$$\bar{h}_{j} * y_{N}[n] = \frac{1}{\sqrt{N}} \sum_{n} \bar{h}[n] = \frac{1}{\sqrt{N}} \sum_{n} h[n] = \sqrt{2}y_{N}[n]$$
(A.13)

and

$$\bar{g}_j * y_N[n] = \frac{1}{\sqrt{N}} \sum_n \bar{g}[n] = \frac{1}{\sqrt{N}} \sum_n g[n] = 0.$$
 (A.14)

From (A.13) and (A.14) we get that $\mathbf{a}_1 = \sqrt{2}\mathbf{a}_0 = \sqrt{2}\mathbf{y}_N$ and $\mathbf{d}_1 = \mathbf{0}$, where $\mathbf{0}$ denotes a vector of all zeros. Using similar calculations it can be shown that $\mathbf{a}_\ell = (\sqrt{2})^\ell \mathbf{y}_N$ and $\mathbf{d}_\ell = \mathbf{0}$, therefore $\|\mathbf{c}\|^2 = 2^L \|\mathbf{y}_N\|^2$. Now, let be $\tilde{\mathbf{y}}_N$ a vector which satisfies $\tilde{y}_N[k] = \frac{1}{\sqrt{N}}(-1)^k$, k = 1..., N. Then using a similar procedure it can be shown that the vector $\tilde{\mathbf{P}}_0 \tilde{\mathbf{y}}_N$ meets the lower bound.

B. Proof of Proposition 2

We show that the representation \mathbf{c}^{SA} is a frame by noticing that $\|\mathbf{c}^{SA}\|^2 = \sum_n |\langle \phi_n^{SA}, \mathbf{y} \rangle|^2$, and showing that

$$2\|\mathbf{y}\|^{2} \le \|\mathbf{c}^{SA}\|^{2} \le 2^{L}n\|\mathbf{y}\|^{2}.$$
 (A.15)

We first use (15) and see that

$$\|\mathbf{c}^{SA}\|^{2} = \mathbf{y}^{T} \left(\Phi^{SA}\right)^{T} \Phi^{SA} \mathbf{y} = \sum_{j=1}^{n} \mathbf{y}^{T} \mathbf{R}_{j}^{T} \Phi^{T} \Phi \mathbf{R}_{j} \mathbf{y}$$
$$= \sum_{j=1}^{n} \|\Phi \mathbf{R}_{j} \mathbf{y}\|^{2}$$
(A.16)

From Proposition 1 we have that

$$2\|\mathbf{R}_{j}\mathbf{y}\|^{2} \leq \|\Phi\mathbf{R}_{j}\mathbf{y}\|^{2} \leq 2^{L}\|\mathbf{R}_{j}\mathbf{y}\|^{2}.$$
 (A.17)

We notice that

$$\sum_{j=1}^{n} \|\mathbf{R}_{j}\mathbf{y}\|^{2} = \mathbf{y}^{T} \left(\sum_{j=1}^{n} \mathbf{R}_{j}^{T} \mathbf{R}_{j} \right) \mathbf{y} = \mathbf{y}^{T} \mathbf{D} \mathbf{y}$$
(A.18)

and therefore

$$2\mathbf{y}^T \mathbf{D} \mathbf{y} \le \|\mathbf{c}^{SA}\|^2 \le 2^L \mathbf{y}^T \mathbf{D} \mathbf{y}.$$
 (A.19)

As the minimum value of the diagonal matrix \mathbf{D} equals 1, and the maximum value equals n, we obtain that

$$2\|\mathbf{y}\|^{2} \le \|\mathbf{c}^{SA}\|^{2} \le 2^{L}n\|\mathbf{y}\|^{2}.$$
 (A.20)

ACKNOWLEDGEMENTS

The authors would like to thank the anonymous reviewers for their helpful comments.

REFERENCES

- I. Ram, M. Elad, and I. Cohen, "Redundant wavelets on graphs and high dimensional data clouds," *IEEE Signal Process. Lett.*, vol. 19, no. 5, pp. 291–294, May 2012.
- [2] A. Buades, B. Coll, and J. M. Morel, "A review of image denoising algorithms, with a new one," *Multiscale Model. Simul.*, vol. 4, no. 2, pp. 490–530, 2006.
- [3] I. Ram, M. Elad, and I. Cohen, "Image processing using smooth ordering of its patches," *IEEE Trans. Image Process.*, vol. 22, no. 7, pp. 2764–2774, Jul. 2013.
- [4] M. Elad, J. L. Starck, P. Querre, and D. L. Donoho, "Simultaneous cartoon and texture image inpainting using morphological component analysis (MCA)," *Appl. Comput. Harmon. Anal.*, vol. 19, no. 3, pp. 340–358, 2005.
- [5] G. Plonka, "The easy path wavelet transform: A new adaptive wavelet transform for sparse representation of two-dimensional data," *Multiscale Model. Simul.*, vol. 7, no. 3, pp. 1474–1496, 2009.
- [6] M. Elad and M. Aharon, "Image denoising via sparse and redundant representations over learned dictionaries," *IEEE Trans. Image Process.*, vol. 15, no. 12, pp. 3736–3745, Dec. 2006.
- [7] J. Mairal, F. Bach, J. Ponce, G. Sapiro, and A. Zisserman, "Non-local sparse models for image restoration," in *Proc. IEEE 12th Int. Conf. Comput. Vision.* Sep./Oct. 2009, pp. 2272–2279.
- [8] G. Yu, G. Sapiro, and S. Mallat, "Solving inverse problems with piecewise linear estimators: From Gaussian mixture models to structured sparsity," *IEEE Trans. Image Process.*, vol. 21, no. 5, pp. 2481–2499, May 2012.
- [9] K. Dabov, A. Foi, V. Katkovnik, and K. Egiazarian, "Image denoising by sparse 3D transform-domain collaborative filtering," *IEEE Trans. Image Process.*, vol. 16, no. 8, pp. 2080–2095, Aug. 2007.
- [10] K. Dabov, A. Foi, V. Katkovnik, and K. Egiazarian, "Image restoration by sparse 3D transform-domain collaborative filtering," *Proc. SPIE Electron. Imag.*, vol. 6812, Mar. 2008.
- [11] A. Danielyan, V. Katkovnik, and K. Egiazarian, "BM3D frames and variational image deblurring," *IEEE Trans. Image Process.*, vol. 21, no. 4, pp. 1715–1728, Apr. 2012.

- [12] I. Ram, M. Elad, and I. Cohen, "Generalized tree-based wavelet transform," *IEEE Trans. Signal Process.*, vol. 59, no. 9, pp. 4199–4209, Sep. 2011.
- [13] M. Shensa, "The discrete wavelet transform: Wedding the a trous and Mallat algorithms," *IEEE Trans. Signal Process.*, vol. 40, no. 10, pp. 2464–2482, Oct. 1992.
- [14] G. Beylkin, "On the representation of operators in bases of compactly supported wavelets," *SIAM J. Numer. Anal.*, vol. 29, no. 6, pp. 1716–1740, 1992.
- [15] T. H. Cormen, Introduction to Algorithms. Cambridge, MA, USA: MIT Press, 2001.
- [16] S. Mallat, A Wavelet Tour of Signal Processing: The Sparse Way. San Francisco, CA, USA: Academic, 2009.
- [17] M. Holschneider, R. Kronland-Martinet, J. Morlet, and P. Tchamitchian, "A real-time algorithm for signal analysis with the help of the wavelet transform," in *Wavelets: Time-Frequency Methods and Phase Space*, vol. 1. Berlin, Germany: Springer-Verlag, 1989, pp. 286–297.
- [18] J. E. Fowler, "The redundant discrete wavelet transform and additive noise," *IEEE Signal Process. Lett.*, vol. 12, no. 9, pp. 629–632, Sep. 2005.
- [19] M. Elad, P. Milanfar, and R. Rubinstein, "Analysis versus synthesis in signal priors," *Inverse Problems*, vol. 23, no. 3, pp. 947–968, 2007.
- [20] S. P. Ghael, A. M. Sayeed, and R. G. Baraniuk, "Improved wavelet denoising via empirical Wiener filtering," *Proc. SPIE*, vol. 3169, Oct. 1997.



Israel Cohen (M'01–SM'03) is a Professor of Electrical Engineering with Technion–Israel Institute of Technology, Haifa, Israel, where he received the B.Sc. (*summa cum laude*), M.Sc., and Ph.D. degrees in electrical engineering in 1990, 1993, and 1998, respectively. From 1990 to 1998, he was a Research Scientist with Rafael Research Laboratories, Israel Ministry of Defense, Haifa. From 1998 to 2001, he was a Post-Doctoral Research Associate with the Computer Science Department, Yale University, New Haven, CT, USA. In 2001, he joined

the Electrical Engineering Department, Technion. His research interests are statistical signal processing, analysis and modeling of acoustic signals, speech enhancement, noise estimation, microphone arrays, source localization, blind source separation, system identification, and adaptive filtering. He is a Co-Editor of the Multichannel Speech Processing Section of the Springer Handbook of Speech Processing (Springer, 2008) and Speech Processing in Modern Communication: Challenges and Perspectives (Springer, 2010), has co-authored Noise Reduction in Speech Processing (Springer, 2009), and a General Co-Chair of the 2010 International Workshop on Acoustic Echo and Noise Control. He was a recipient of the Alexander Goldberg Prize for Excellence in Research, and the Muriel and David Jacknow Award for Excellence in Teaching. He serves as a member of the IEEE Audio, Acoustic Signal Processing Technical Committee, and the IEEE Speech and Language Processing Technical Committee. He served as an Associate Editor of the IEEE TRANSACTIONS ON AUDIO, SPEECH, AND LANGUAGE PROCESSING and the IEEE SIGNAL PROCESSING LETTERS, and as Guest Editor of a Special Issue of the EURASIP Journal on Advances in Signal Processing on Advances in Multimicrophone Speech Processing and a Special Issue of the Elsevier Speech Communication Journal on Speech Enhancement.



Idan Ram received the B.Sc. (*cum laude*) and M.Sc. degrees in electrical engineering from Technion–Israel Institute of Technology, Haifa, Israel, in 2004 and 2009, respectively, where he is currently pursuing the Ph.D. degree in electrical engineering. He was a recipient of the Ollendorff Award for research in the field of image processing and analysis in 2013. His research interests are image processing using sparse signal representations and smooth patch ordering.



Michael Elad (F'12) received the B.Sc., M.Sc., and D.Sc. degrees from the Department of Electrical engineering from Technion–Israel Institute of Technology, Haifa, Israel, in 1986, 1988, and 1997, respectively, where he has been a Faculty Member with the Department of Computer Science, since 2003, and a Full-Professor since 2010. He is involved in signal and image processing, specializing, in particular, on inverse problems, sparse representations and superresolution. He was a recipient of the Technion's Best Lecturer Award six times,

the 2007 Solomon Simon Mani Award for excellence in teaching, the 2008 Henri Taub Prize for academic excellence, and the 2010 Hershel-Rich Prize for innovation. He is serving as an Associate Editor of the *SIAM Journal on Imaging Sciences* and *Applied and Computational Harmonic Analysis*, and a Senior Editor of the IEEE SIGNAL PROCESSING LETTERS.